

Dynamic Complex-Frequency Control of Grid-Forming Converters



Swiss Federal Inst. of Tech. (ETH), Zurich, Switzerland

4 Nov 2024

ETH zürich



Grid-forming inverters are crucial!









Droop control and VSM

 $\dot{\theta} - \omega_0 = \eta (p^* - p)$ or $T_J \ddot{\theta} + D(\dot{\theta} - \omega_0) = p^* - p$ $\dot{v} = \eta \alpha (v^* - v) + \eta (q^* - q)$

- Single-input single-output: based on decoupling and linearization
- Good power-sharing property and smallsignal performance
- Poor large-signal performance

 \rightarrow Need multivariate/MIMO controllers

 $p \rightarrow Droop/ \qquad \theta \qquad Voltage \qquad Current \\ q \rightarrow VSM \qquad Or ontrol \qquad Control \qquad Or ontrol \quad Or ontrol \qquad Or ontrol \quad Or ontr$

$$v_g = 0.1, z_g = 0.4 + j0.4, p^* = 0, q^* = 0, v^* = 1, \alpha = 1, \eta = 0.08$$













- 1. Complex-Frequency Control
- 2. Dynamic Complex-Frequency Control
- 3. Case Studies
- 4. Conclusion









1. Complex-Frequency Control

- 2. Dynamic Complex-Frequency Control
- 3. Case Studies
- 4. Conclusion





5

Complex-frequency control



Complex variables are multivariate

- Complex voltage $\underline{v} = ve^{j\theta}$
- Complex angle $\underline{\vartheta} \coloneqq \ln v + j\theta$

Complex droop control are multivariate

- Multivariate feedback: complex power
- Multivariate control output: complex angle

$$\underline{\dot{\vartheta}} = j\omega_0 + \eta e^{j\varphi} \left(\frac{p^* - jq^*}{v^{*2}} - \frac{p - jq}{v^2} \right) + \eta \alpha \frac{v^{*2} - v^2}{v^{*2}}$$

VS.

SISO droop control: $\dot{\theta} = \omega_0 + \eta \left(p_{\varphi}^{\star} - p_{\varphi} \right)$

- Complex power $\underline{s} = p + jq$ - Complex frequency $\underline{\varpi} \coloneqq \underline{\dot{\vartheta}} = \frac{\dot{v}}{v} + j\dot{\theta}$







Comparison with classics



	Classical droop control	Complex droop control
Controller	$\begin{split} \dot{v} &= \eta \left(q^{\star} - q \right) + \eta \alpha \left(v^{\star} - v \right), \\ \dot{\theta} &= \omega_0 + \eta \left(p^{\star} - p \right). \end{split}$	$\frac{\dot{v}}{v} = \eta \left(\frac{q^{\star}}{v^{\star 2}} - \frac{q}{v^2}\right) + \eta \alpha \frac{v^{\star} - v}{v^{\star}}, \text{ or } + \eta \alpha \frac{v^{\star 2} - v^2}{v^{\star 2}},$ $\dot{\theta} = \omega_0 + \eta \left(\frac{p^{\star}}{v^{\star 2}} - \frac{p}{v^2}\right).$
Stability	Small-signal stability is guaranteed but transient stability not guaranteed (far away from the nominal)	Both small-signal stability and transient stability are theoretically guaranteed
Dynamic performance	Dynamic droop control: VSM and their variants	Dynamic complex-frequency droop control
Transient performance	Virtual admittance + limiter Cross-forming control Threshold virtual impedance	Virtual admittance + limiter Cross-forming control Threshold virtual impedance control
Steady-state performance	Power sharing	Power sharing









1. Complex-Frequency Control

2. Dynamic Complex-Frequency Control

- 3. Case Studies
- 4. Conclusion





Dynamic complex-frequency control



- Static droop
 - Static gains
 - No inertia response

$$\dot{\underline{\vartheta}} = j\omega_0 + \eta e^{j\varphi} \left(\frac{p^* - jq^*}{v^{*2}} - \frac{p - jq}{v^2} \right) + \eta \alpha \frac{v^{*2} - v^2}{v^{*2}}$$

$$\begin{bmatrix} \Delta \varepsilon \\ \Delta \omega \end{bmatrix} = \begin{bmatrix} \mathbf{0} & -\eta \\ \eta & \mathbf{0} \end{bmatrix} \left(\begin{bmatrix} -\Delta \rho \\ \Delta \sigma \end{bmatrix} - \begin{bmatrix} \mathbf{0} \\ -\alpha \end{bmatrix} \frac{\Delta v}{v^*} \right)$$

Dynamic droop

- Dynamic gains (Transfer Fcns)
- Rich dynamic responses

$$\begin{bmatrix} \Delta \varepsilon \\ \Delta \omega \end{bmatrix} = \begin{bmatrix} T^{\text{re}}(s) & -T^{\text{im}}(s) \\ T^{\text{im}}(s) & T^{\text{re}}(s) \end{bmatrix} \left(\begin{bmatrix} -\Delta \rho \\ \Delta \sigma \end{bmatrix} - \begin{bmatrix} T^{\nu p}(s) \\ T^{\nu q}(s) \end{bmatrix} \frac{\Delta \nu}{\nu^*} \right)$$
$$:= \underline{T}(s) \qquad := \underline{T}^{\nu}(s)$$

$$\Delta \underline{\overline{\varpi}} = \underline{T}(s) \left(-\Delta \underline{\varsigma} - \underline{T}^{\nu}(s) \Delta \nu_{\text{pcc}} \right)$$





Single-converter control setup





$$\Delta \underline{\varpi} = \underline{T}(s) \left(-\Delta \underline{\varsigma} - \underline{T}^{\nu}(s) \Delta \nu_{\text{pcc}} \right)$$

We choose $\underline{T}(s) = \underline{T}_{des}(s)$ and $\underline{T}^{\nu}(s) = \underline{T}^{\nu}_{des}(s)$ to provide some **desired dynamic behavior**







Multi-converter control setup (Dynamic VPP)





$$\Delta \underline{\varpi}_{\rm pcc} = \underline{T}_{\rm des}(s) \left(\Delta \underline{\varsigma}_{\rm pcc} - \underline{T}_{\rm des}^{\nu}(s) \Delta v_{\rm pcc} \right)$$

Aggregation conditions for complex-frequency control

$$\left(\sum_{k} \underline{T}_{k}(s)^{-1}\right)^{-1} \stackrel{!}{=} \underline{T}_{des}(s) \qquad \sum_{k} \underline{T}_{k}^{\nu}(s) \stackrel{!}{=} \underline{T}_{des}^{\nu}(s)$$

$$\Delta \underline{\varpi}_{\rm pcc} = \left(\sum_{k} \underline{T}_{k}(s)^{-1} \right)^{-1} \left(\Delta \underline{\varsigma}_{\rm pcc} - \sum_{k} \underline{T}_{k}^{\nu}(s) \, \Delta \nu_{\rm pcc} \right)$$



Dynamic Complex-Frequency Control of Grid-Forming Converters – Xiuqiang He (ETH Zurich)

 \approx



Industrial Electronics





- 1. Complex-Frequency Control
- 2. Dynamic Complex-Frequency Control
- 3. Case Studies
- 4. Conclusion





Case study setup









Control structure

Society





Case study results





Dynamic Complex-Frequency Control of Grid-Forming Converters – Xiuqiang He (ETH Zurich)

ETH zürich 15





- 1. Complex-Frequency Control
- 2. Dynamic Complex-Frequency Control
- 3. Case Studies
- 4. Conclusion





Conclusion / Take-home messages



- Complex-frequency (MIMO) grid-forming control: It outperforms classical (SISO) ones in stability guarantees.
- **Dynamic complex-frequency control: Transfer functions** (dynamic gains) instead of static gains provide richer dynamic responses.
- **DVPPs, dynamic virtual power plants:** A good solution to coordinate multiple converters to fulfill a desired response.
- Future work: Stability guarantees when using dynamic (vs. static) complexfrequency control.





Further reading



Complex frequency

• F. Milano, <u>Complex frequency</u>, IEEE TPWRS, 2022.

Stability guarentees of complex droop control (aka. dVOC)

- M Colombino, D Groß, JS Brouillon, F Dörfler, <u>Global phase and magnitude synchronization of coupled</u> oscillators with application to the control of grid-forming power inverters, IEEE TAC, 2020.
- X. He, V. Häberle, and F. Dörfler, <u>Complex-frequency synchronization of converter-based power systems</u>, IEEE TCNS, 2022.
- X. He, L. Huang, I. Subotić, V. Häberle, and F. Dörfler, <u>Quantitative stability conditions for grid-forming</u> <u>converters with complex droop control</u>, IEEE TPEL, 2024.
- X. He, V. Häberle, I. Subotić, and F. Dörfler, <u>Nonlinear stability of complex droop control in converter-based</u> power systems, IEEE L-CSS, 2023.
- X. He and F. Dörfler, <u>Passivity and decentralized stability conditions for grid-forming converters</u>, IEEE TPWRS, 2024.





Acknowledgments



Thank you!

Dynamic Complex-Frequency Control of Grid-Forming Converters

Xiuqiang He (何秀强), Senior Scientist

ETH Zurich, Switzerland 4 Nov, 2024 This work was supported by the European Union's Horizon 2020 and 2023 research and innovation programs (Grant Agreement Numbers 883985 and 101096197).







